

E 2.5 Signals & Linear Systems. ①

Tutorial Sheet 5 - Solutions

1). a). $\mathcal{L}\{u(t)\} = \frac{1}{s}$ $\mathcal{L}\left\{\frac{d}{dt}\right\} = s$.
 Given the initial condition,

$$(s^2 + 3s + 2) Y(s) = s \left(\frac{1}{s}\right)$$

$$\Rightarrow Y(s) = \frac{1}{s^2 + 3s + 2} = \frac{1}{s+1} - \frac{1}{s+2}$$

$$\therefore y(t) = (e^{-t} - e^{-2t}) u(t) //$$

b) $(s^2 Y(s) - 2s - 1) + 4(s Y(s) - 2) + 4Y(s)$
 $= (s+1) \times \frac{1}{s+1}$

$$\Rightarrow (s^2 + 4s + 4) Y(s) = 2s + 10$$

$$\Rightarrow Y(s) = \frac{2s + 10}{s^2 + 4s + 4} = \frac{2s + 10}{(s+2)^2} = \frac{2}{s+2} + \frac{4}{(s+2)^2}$$

$$\therefore y(t) = (2 + 6t) e^{-2t} u(t) //$$

c) $(s^2 Y(s) - s - 1) + 6(s Y(s) - 1) + 25 Y(s)$
 $= (s+2) \frac{25}{s} = 25 + \frac{50}{s}$

$$\Rightarrow (s^2 + 6s + 25) Y(s) = \frac{s^2 + 32s + 50}{s}$$

$$\Rightarrow Y(s) = \frac{s^2 + 32s + 50}{s(s^2 + 6s + 25)} = \frac{2}{s} + \frac{-s + 20}{s^2 + 6s + 25}$$

$$\therefore y(t) = \left[2 + 5.836 e^{-3t} \cos(4t - 99.86^\circ) \right] u(t) //$$

2) a) $Y(s) = \frac{5s + 3}{s^2 + 11s + 24}$ // (Write down directly!) (2)

b) $Y(s) = \frac{3s^2 + 7s + 5}{s^3 + 6s^2 + 11s + 6}$ //

c) $Y(s) = \frac{3s + 2}{s^4 + 4s}$
 $= \frac{3s + 2}{s(s^3 + 4)}$ //

3) a) i) $f(t) = e^{-4t} u(t) \therefore F(s) = \frac{1}{s+4}$

$Y(s) = H(s) F(s) = \frac{s+5}{(s+2)(s+3)(s+4)}$
 $= \frac{3/2}{s+2} - \frac{2}{s+3} + \frac{1/2}{s+4}$

$\therefore y(t) = \left[\frac{3}{2} e^{-2t} - 2e^{-3t} + \frac{1}{2} e^{-4t} \right] u(t)$ //
 (Easy with LTP)

ii) $f(t) = e^{-3t} u(t) \therefore F(s) = \frac{1}{s+3}$

$Y(s) = \frac{s+5}{(s+2)(s+3)^2} = \frac{3}{s+2} - \frac{3}{s+3} - \frac{2}{(s+3)^2}$

$\therefore y(t) = (3e^{-2t} - 3e^{-3t} - 2te^{-3t}) u(t)$ //

$$\text{iii) } f(t) = e^{-4(t-5)} u(t-5) \quad (3)$$

~~$\therefore F(s) =$~~

Note that this is delayed by 5 sec.

$$\therefore F(s) = \left(\frac{1}{s+4} \right) e^{-5s} \quad \left(\begin{array}{l} \text{Time} \\ \text{shifting} \\ \text{property} \end{array} \right)$$

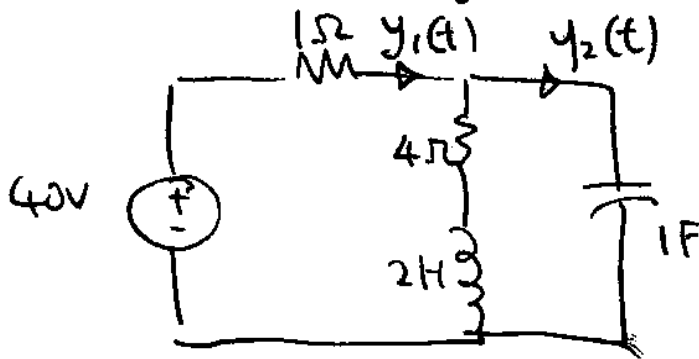
$$\begin{aligned} Y(s) &= \frac{s+5}{(s+2)(s+3)(s+4)} e^{-5s} \\ &= \left[\frac{3/2}{s+2} - \frac{2}{s+3} + \frac{1/2}{s+4} \right] e^{-5s} \end{aligned}$$

$$\therefore y(t) = \left[\frac{3}{2} e^{-2(t-5)} - 2 e^{-3(t-5)} + \frac{1}{2} e^{-4(t-5)} \right] u(t-5)$$

$$\text{b) } \frac{d^2 y}{dt^2} + 2 \frac{dy}{dt} + 5y(t) = 2 \frac{df}{dt} + 3f(t)$$

4. At $t=0$, inductor current $y_1(0) = 4A$.
 Capacitor voltage is 16V.

(4)



1. After $t=0$, loop equations are: - (Voltage around loops)

$$2\left(\frac{dy_1}{dt} - \frac{dy_2}{dt}\right) + y_1(t) + 4(y_1(t) - y_2(t)) = 40$$

$$\Rightarrow 2\frac{dy_1}{dt} - 2\frac{dy_2}{dt} + 5y_1(t) - 4y_2(t) = 40 \quad (1)$$

and

$$-2\frac{dy_1}{dt} - 4y_1(t) + 2\frac{dy_2}{dt} + 4y_2(t) + \int_{-\infty}^t y_2(\tau) d\tau = 0 \quad (2)$$

Initial conditions gives us:

$$y_1(t) \Leftrightarrow Y_1(s), \quad \frac{dy_1}{dt} = sY_1(s) - 4$$

$$y_2(t) \Leftrightarrow Y_2(s), \quad \frac{dy_2}{dt} = sY_2(s)$$

$$\int_{-\infty}^t y_2(\tau) d\tau \Leftrightarrow \frac{1}{s}Y_2(s) + \frac{16}{s}$$

Laplace Transform of the loop equations are therefore:

$$2(sY_1(s) - 4) - 2sY_2(s) + 5Y_1(s) - 4Y_2(s) = \frac{40}{s}$$

$$\Rightarrow (2s+5)Y_1(s) - (2s+4)Y_2(s) = 8 + \frac{40}{s} \quad (1s)$$

~~$(2s+5)Y_1(s) - (2s+4)Y_2(s) = 8 + \frac{40}{s}$~~

$$-2(sY_1(s) - 4) - 4Y_1(s) + 2sY_2(s) + 4Y_2(s) + \frac{1}{s}Y_2(s) + \frac{16}{s} = 0$$

$$\Rightarrow -(2s+4)Y_1(s) + (2s+4 + \frac{1}{s})Y_2(s) = -8 - \frac{16}{s} \quad (2s)$$

Write in matrix form:

(5)

$$\begin{bmatrix} (2s+5) & -(2s+4) \\ -(2s+4) & (2s+4+\frac{1}{s}) \end{bmatrix} \begin{bmatrix} Y_1(s) \\ Y_2(s) \end{bmatrix} = \begin{bmatrix} 8 + \frac{40}{s} \\ -8 - \frac{16}{s} \end{bmatrix}$$

Apply Cramer's rule yields

$$Y_1(s) = \frac{4(6s^2 + 13s + 5)}{s(s^2 + 3s + 2.5)} = \frac{8}{s} + \frac{16s + 28}{s^2 + 3s + 2.5}$$

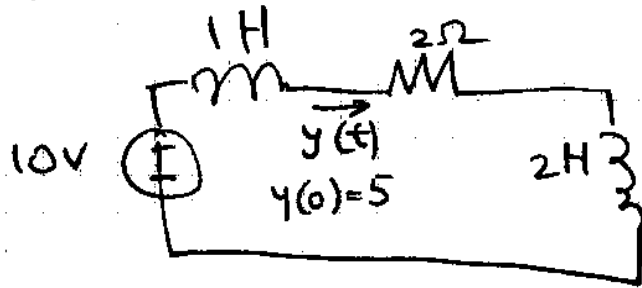
Use table:

$$\therefore y_1(t) = \left[8 + 17.89 e^{-1.5t} \cos\left(\frac{t}{2} - 26.56^\circ\right) \right] u(t)$$

$$Y_2(s) = \frac{20(s+2)}{(s^2 + 3s + 2.5)}$$

$$\therefore y_2(t) = 20\sqrt{2} e^{-1.5t} \cos\left(\frac{t}{2} - \frac{\pi}{4}\right) u(t)$$

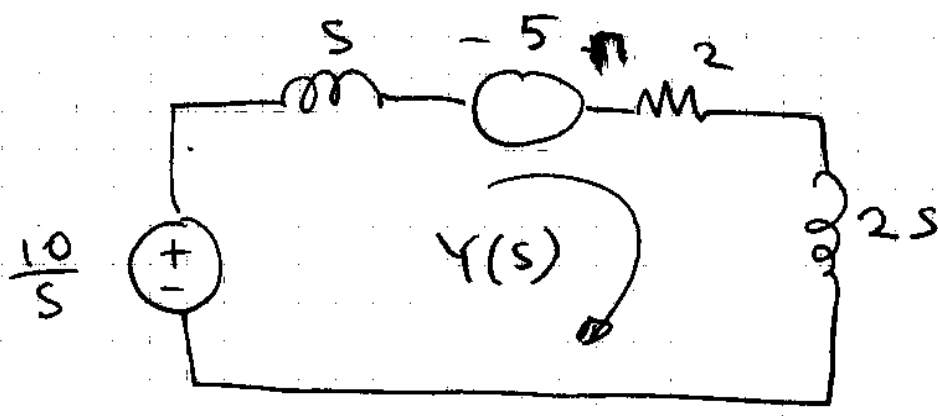
5. Before switch is opened, (ie $t=0$) inductor current is 5A, ie. $y(0)=5$.



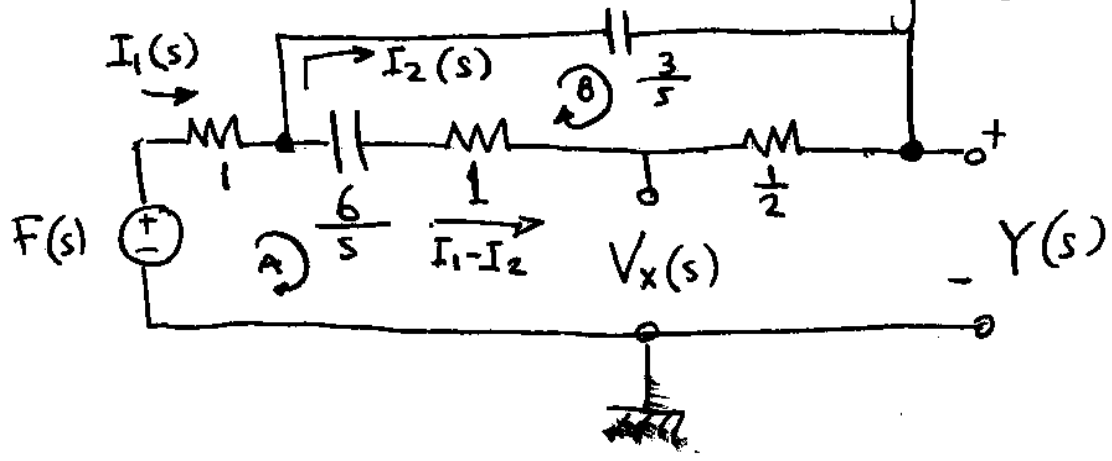
$$\begin{aligned}
 Y(s) &= \frac{10/s + 5}{3s + 2} \\
 &= \frac{5s + 10}{s(3s + 2)} \\
 &= \frac{5}{3} \left[\frac{3}{s} - \frac{2}{s + \frac{2}{3}} \right]
 \end{aligned}$$

$$\therefore y(t) = \left(5 - \frac{10}{3} e^{-\frac{2}{3}t} \right) u(t)$$

Transformed circuit in s-domain:



6. The transformed circuit for Fig Q6 :



Since op amp has infinite gain,
 $V_x(s) \approx 0$.

loop equations are: —

(A) $1 \times I_1(s) + \left(\frac{6}{s} + 1\right) [I_1(s) - I_2(s)] = F(s)$

(B) $-\frac{3}{2} I_1(s) + \left(\frac{6}{s} + \frac{3}{2}\right) [I_1(s) - I_2(s)] = 0$

$$\begin{bmatrix} \left(\frac{6}{s} + 2\right) & -\left(\frac{6}{s} + 1\right) \\ \frac{6}{s} & \left(\frac{6}{s} + \frac{3}{2}\right) \end{bmatrix} \begin{bmatrix} I_1(s) \\ I_2(s) \end{bmatrix} = \begin{bmatrix} F(s) \\ 0 \end{bmatrix}$$

Cramer's rule yields

$$I_1(s) = \frac{s(s+6)}{s^2 + 8s + 12} F(s), \quad I_2(s) = \frac{s(s+4)}{s^2 + 8s + 12} F(s)$$

$$Y(s) = -\frac{1}{2} \times [I_1(s) - I_2(s)]$$

$$= \frac{-s}{s^2 + 8s + 12} F(s)$$

∴ Transfer Function

$$H(s) = \frac{Y(s)}{F(s)} = -\frac{s}{s^2 + 8s + 12} //$$

8

$$u(t) \Leftrightarrow \frac{1}{s}$$

$$7)(a) \quad Y(s) = \frac{6s^2 + 3s + 10}{s(2s^2 + 6s + 5)}$$

$$\therefore y(0^+) = \lim_{s \rightarrow \infty} s Y(s) = 3$$

$$y(\infty) = \lim_{s \rightarrow 0} s Y(s) = 2 \quad //$$

$$(b) \quad e^{-t} u(t) \Leftrightarrow \frac{1}{s+1}$$

$$\therefore Y(s) = \frac{6s^2 + 3s + 10}{(s+1)(2s^2 + 6s + 5)}$$

$$\therefore y(0^+) = \lim_{s \rightarrow \infty} s Y(s) = 3$$

$$y(\infty) = \lim_{s \rightarrow 0} s Y(s) = 0 \quad //$$

$$8. \quad H(s) = \frac{s+3}{(s+2)^2}$$

$$\therefore H(j\omega) = \frac{j\omega + 3}{(j\omega + 2)^2}$$

$$|H(j\omega)| = \frac{\sqrt{\omega^2 + 9}}{\omega^2 + 4}$$

$$\angle H(j\omega) = \tan^{-1}\left(\frac{\omega}{3}\right) - \tan^{-1}\left(\frac{\omega}{2}\right) \times 2$$

$$(a) \quad f(t) = \cos(2t + 60^\circ) \quad \therefore \omega = 2$$

$$|H(j2)| = \frac{\sqrt{13}}{8} \quad \text{and} \quad \angle H(j2) = 33.69^\circ - 2 \times 45^\circ = -56.31^\circ$$

$$\therefore y(t) = \frac{\sqrt{13}}{8} \cos(2t + 60^\circ - 56.31^\circ)$$

$$= \frac{\sqrt{13}}{8} \cos(2t + 3.69^\circ) \quad //$$

9

8 (b) $f(t) = \sin(3t - 45^\circ)$. Hence $\omega = 3$,

$$|H(j3)| = \frac{\sqrt{18}}{13} \quad \text{and} \quad \angle H(j3) = 45^\circ - 112.62^\circ = -67.62^\circ.$$

$$\therefore y(t) = \frac{\sqrt{18}}{13} \sin(3t - 112.62^\circ) //$$

(c) $f(t) = e^{j3t}$ $\omega = 3$

$$\begin{aligned} y(t) &= H(j\omega) e^{j3t} = H(j3) e^{j3t} \\ &= |H(j3)| e^{j(3t + \angle H(j3))} \quad \text{From (b)} \end{aligned}$$

$$= \frac{\sqrt{18}}{13} \sin(3t - 67.62^\circ) //$$

10. a) If r and d are the distance of the zero and pole from $j\omega$ respectively, $|H(j\omega)|$ is the ratio $\frac{r}{d}$ corresponding to $j\omega$.

At $\omega=0$, $\frac{r}{d} = 0.5$
 $\omega=\infty$, $\frac{r}{d} = 1$

~~$\angle H(j\omega)$~~
At $\omega=0$, $\angle H(j\omega) = 0$
 $\omega \rightarrow \infty$, $\angle H(j\omega) = 0$ } In between, the angle is +ve.

